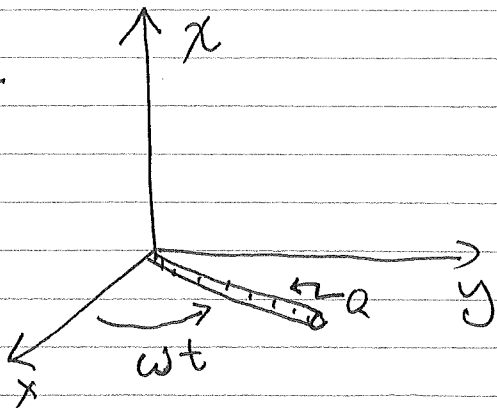


# Physics 406: E&M II

## Problem Set #10: Solutions

Prob. 1



A rotating rod has a line charge  $Q$  uniformly distributed over its length. It rotates in the  $x$ - $y$  plane at angular velocity  $\omega$ , with  $\omega \ll \frac{c}{L}$ .

Because the motion is slow, we can make the multipole expansion under the small parameter  $\frac{\omega}{c} L = \beta L \ll 1$ .

$\Rightarrow$  The dominant power radiated is electric dipole radiation.

(a) We need to calculate the electric dipole moment, which will oscillate @ freq.  $\omega$

$$\vec{p}(t) = \text{Re}(\vec{p} e^{-i\omega t})$$

By definition  $L$

$$\begin{aligned} \vec{p}(t) &= \int_0^L ds \frac{Q}{L} s (\cos \omega t \hat{x} + \sin \omega t \hat{y}) \\ &= \frac{Q}{2L} s^2 \Big|_0^L \text{Re}[(\hat{x} + i\hat{y}) e^{-i\omega t}] \\ &= \frac{QL}{2} \end{aligned}$$

$$\Rightarrow \vec{p} = \frac{QL}{2} (\hat{x} + i\hat{y})$$

electric dipole

The time averaged  $\hat{P}$  power radiated in electric dipole radiation is given by

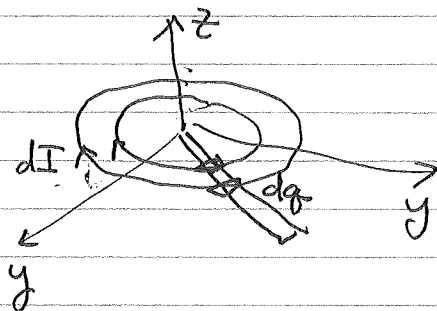
$$\langle P_{\text{rad}}^{\text{E1}} \rangle = \left( \frac{1}{4\pi\epsilon_0} \right) \frac{ck^4 |\vec{p}|^2}{3}$$

$$= \frac{1}{4\pi\epsilon_0} \frac{ck^4}{3} \left| \frac{qL}{2} (\ddot{x} + i\dot{y}) \right|^2$$

$$\Rightarrow \boxed{\langle P_{\text{rad}}^{\text{E1}} \rangle = \frac{1}{4\pi\epsilon_0} \frac{ck^4}{6} (qL)^2 \ddot{x}^2}$$

(b) The power radiated as magnetic dipole radiation  
 $P_{\text{rad}} \propto |\ddot{m}|^2$

The key question, thus, is whether there is a magnetic dipole moment that has a ~~finite~~ nonzero second derivative.

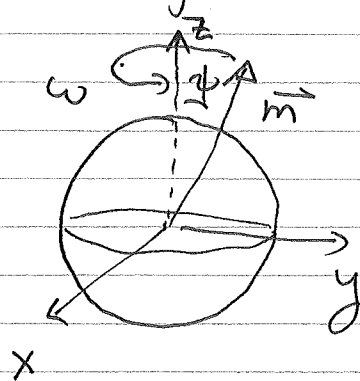


If we break up the rod into elements  $dq$ , each is rotating around the z-axis at constant ~~speed~~ speed  $\omega$ .

Thus, each current loop has ~~constant~~ constant current  $\Rightarrow$  the magnetic dipole is constant

Thus, there is no magnetic dipole radiation

## Problem 2: Magnetic Dipole Radiation



Rotating neutron star  
= Pulsar

$$\vec{m}(t) = |\vec{m}| \left[ \underbrace{\sin \psi (\cos \omega t \hat{x} + \sin \omega t \hat{y})}_{\text{rotating component}} + \cos \psi \hat{z} \right]$$

The rotating component will radiate magnetic dipole radiation at an average rate

$$\langle P_{\text{rad}} \rangle = \frac{1}{4\pi\epsilon_0} \frac{k^4 |\vec{m}_{\text{rot}}|^2}{3c}$$

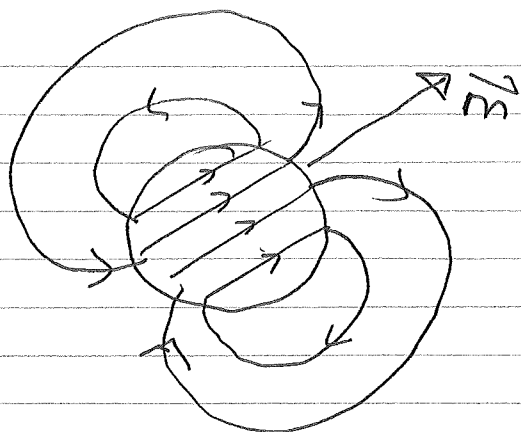
$$\text{Here } \vec{m}_{\text{rot}} = |\vec{m}| \sin \psi (\hat{x} + i\hat{y})$$

$$\Rightarrow \langle P_{\text{rad}} \rangle = \frac{1}{4\pi\epsilon_0} \frac{2k^4 \sin^2 \psi |\vec{m}|^2}{3c}$$

But, what is  $|\vec{m}|$ ? We are given the magnetic field at the surface of the star.

We can think of the neutron star as a uniformly magnetized sphere. In the near field,

We know the field is the same as the (quasi) static  $\vec{B}$ -field



(Near) Field of a uniformly magnetized sphere

From Griffiths, Example 6.1, the magnetic field inside the sphere (up to the radius) is uniform, with the value

$$\vec{B} = \frac{2}{3} \mu_0 \vec{M}$$

$$\vec{M} = \frac{\vec{m}}{\frac{4}{3} \pi R^3} = \text{magnetization}$$

$$\Rightarrow \vec{m} = \frac{4}{3} \pi R^3 \vec{M} = \frac{2\pi}{\mu_0} R^3 \vec{B}$$

$$\begin{aligned} \text{Thus } \langle P_{\text{rad}} \rangle &= \frac{1}{4\pi\epsilon_0} \frac{8\pi^2}{3\mu_0^2} \frac{k^4 R^6 \sin^2 \psi}{c} |\vec{B}|^2 \\ &= \frac{2\pi}{\mu_0} \frac{c k^4 R^6 \sin^2 \psi}{3} |\vec{B}|^2 \end{aligned}$$

Numbers:  $\mu_0 = 4\pi \times 10^{-7} \text{ N/A}$ ,  $\psi = 10^\circ = 0.17 \text{ rad}$

$$k = \frac{\omega}{c}, \quad \omega = \frac{2\pi}{T} = 2\pi \times 10^3 \text{ s}^{-1}$$

$$|\vec{B}| = 10^8 \text{ Tesla}, \quad R = 10^4 \text{ m}$$

$$\Rightarrow \langle P_{\text{rad}} \rangle \approx 3 \times 10^{34} \text{ Watts}$$

Note: 1 year  $\approx 3 \times 10^7$  seconds

$$\Rightarrow \boxed{\langle P_{\text{rad}} \rangle \approx 10^{27} \text{ Joules/year}}$$

The radiated energy comes from the rotational energy of the pulsar

$$\frac{d}{dt} E_{\text{rotation}} \approx - \langle P_{\text{rad}} \rangle$$

$$E_{\text{rotation}} = \frac{2}{5} M_{\text{sun}} R^2 \omega \Rightarrow \frac{dE_{\text{rot}}}{dt} = \frac{4}{5} M_{\text{sun}} R^2 \omega \frac{d\omega}{dt}$$

$$\Rightarrow \boxed{\frac{d\omega}{dt} = - \frac{\langle P_{\text{rad}} \rangle}{\frac{4}{5} M_{\text{sun}} R^2 \omega}}$$

This is a complicated ODE because  $\langle P_{\text{rad}} \rangle$  depends on  $\omega$ .

We can estimate the initial rate of decay

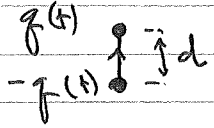
$$\Gamma = \left| \frac{d\omega}{dt} \right|_0 = \frac{\langle P_{\text{rad}} \rangle}{\frac{4}{5} M_{\text{sun}} R^2 (\omega(0))^2} \approx 4.6 \times 10^{-5} \text{ year}^{-1}$$

The lifetime  $\boxed{\tau_{\text{life}} \approx 2 \times 10^4 \text{ years}}$

$$\tau_{\text{life}} \approx \frac{1}{\Gamma}$$

### Problem 3: Radiation by Antennas

(a) The idealized dipole radiator studied in Griffiths and lecture



$$I(t) = \frac{dq}{dt} = -\omega q_0 \sin \omega t = \text{Re}(-i\omega q_0 e^{-i\omega t}) = \text{Re}(\tilde{I}_0 e^{-i\omega t})$$

$$q(t) = q_0 \cos \omega t$$

$$p(t) = q_0 d \cos \omega t$$

The dipole will radiate and carry away energy. This energy must come from a power source that drives the current.

The power lost to radiation appears to the current drive as an effective resistance. In a resistor, energy is lost to heat. Here it is lost to waves, ~~but~~ but the drive can't see the difference.

$$\Rightarrow \underbrace{\frac{1}{2} |\tilde{I}_0|^2 R_{\text{rad}}}_{\text{Power lost in an effective resistor}} = \langle P_{\text{rad}} \rangle$$

Power lost in an effective resistor

$$\frac{1}{2} (\omega q_0)^2 R_{\text{rad}} = \frac{1}{4\pi\epsilon_0} \frac{ck^4}{3} P_0^2 = \frac{1}{4\pi\epsilon_0} \frac{\omega^4 q_0^2 d^2}{3c^3}$$

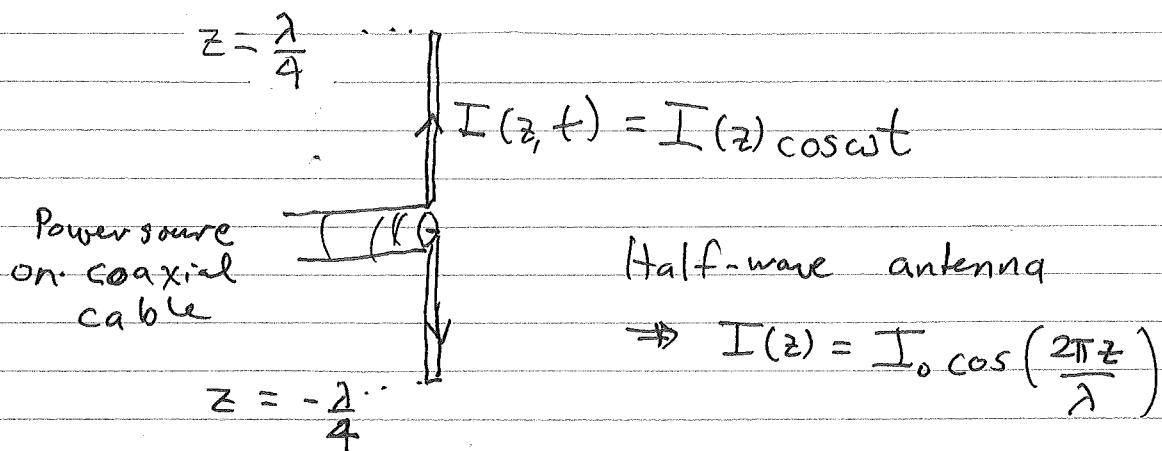
$$\Rightarrow R_{\text{rad}} = \frac{1}{6\pi\epsilon_0} \frac{\omega^2}{c^3} d^2 = \frac{1}{6\pi c \epsilon_0} \left( \frac{2\pi d}{\lambda} \right)^2$$

$\frac{\omega}{c} = 2\pi/\lambda$

$$\Rightarrow \boxed{R_{\text{rad}} = \frac{2}{3} \pi \frac{Z_0}{c} \left( \frac{d}{\lambda} \right)^2 \approx 80 \pi^2 \left( \frac{d}{\lambda} \right)^2 \text{ ohms}}$$

impedance of free space  $\approx 120\pi$  ohms

(b) In a real antenna, the current must go to zero at the ends. We consider here a "center fed" antenna

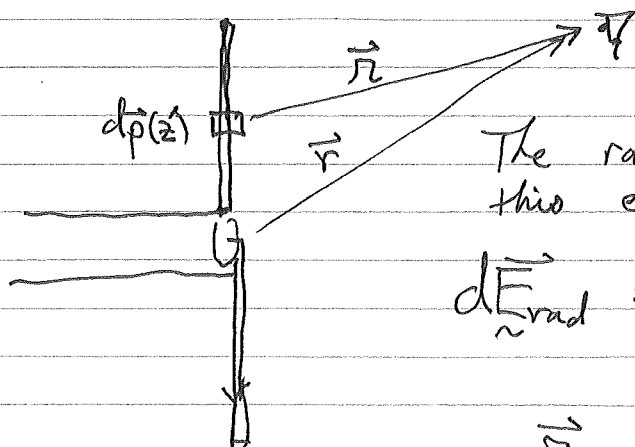


Because the length of the antenna  $l = \frac{\lambda}{2}$  is on the order of  $\lambda$ , we cannot treat this as a single dipole radiator. There are two ways to deal with this:

(i) Break up the antenna into differential elements, each a dipole radiator.

(ii) Go back to first principles, find the potentials, and find the fields.

Consider approach (i) first



The radiation emitted from this element of the antenna

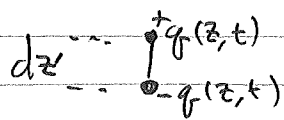
$$d\vec{E}_{\text{rad}} = \frac{1}{4\pi\epsilon_0} k^2 d\vec{p}(z') \frac{e^{ikr}}{r}$$

$$\vec{r} = \vec{r} - z' \hat{z}$$

In the far field,  $R = |\vec{R}| \approx r - \hat{r} \cdot \vec{r}' = r - (\hat{r} \cdot \hat{z}) z'$   
 $= r - \cos\theta z'$

$$\frac{e^{ikR}}{R} \approx \frac{e^{ikr} e^{-ikz' \cos\theta}}{r - \cos\theta z'} \xrightarrow{\text{negligible}}$$

The differential dipole  $d\vec{p}$



$$\frac{dq(z,t)}{dz'} = I(z,t) \Rightarrow -i\omega \tilde{q}(z) = \tilde{I}(z)$$

$$\Rightarrow \tilde{q}(z) = i \frac{\tilde{I}(z)}{\omega}$$

$$\Rightarrow d\vec{p}(z') = \tilde{q}(z') dz' \hat{z} = i \frac{dz'}{\omega} \tilde{I}(z') \hat{z}$$

$$d\vec{p}_{\perp}(z') = i \frac{dz'}{\omega} I(z') \sin\theta \hat{r}$$

$$\Rightarrow d\vec{E}_{\text{rad}} = \frac{1}{4\pi\epsilon_0} \frac{i\omega}{c^2} \frac{e^{ikr}}{r} I(z') e^{-ikz' \cos\theta} \sin\theta \hat{r}$$

The total field is then the integral over the antenna:

$$\vec{E}_{\text{rad}} = \frac{1}{4\pi\epsilon_0} \frac{ik}{c} \frac{e^{ikr}}{r} \sin\theta \int_{-\lambda/4}^{\lambda/4} I(z') e^{-ikz' \cos\theta} dz' \hat{r}$$

$$|\vec{E}_{\text{rad}}|$$



(ii) Method - Find potentials, then  $\vec{E}$

We know:  $\vec{E}(\vec{r}, t) = -\vec{\nabla}V - \frac{\partial \vec{A}(\vec{r}, t)}{\partial t}$

For monochromatic fields:  $\vec{E}(\vec{r}) = -\vec{\nabla}V + i\omega \vec{A}(\vec{r})$

In the far field the effect of the scalar potential is to remove the longitudinal component

$$\Rightarrow \boxed{\vec{E}_{\text{rad}} = i\omega \vec{A}_{\perp}^{\text{rad}}(\vec{r})}$$

Thus, all we have to find is the vector potential

Recall

$$\vec{A}(\vec{r}, t) = \frac{\mu_0}{4\pi} \int d^3r' \frac{\vec{J}(\vec{r}', t_{\text{ret}})}{r}$$

where  $\vec{r} = \vec{r} - \vec{r}'$   
 $t_{\text{ret}} = t - \frac{r}{c}$

Here the current is along a line

$$\Rightarrow \vec{A}(\vec{r}, t) = \frac{\mu_0}{4\pi} \int_{-\lambda/4}^{\lambda/4} dz' \frac{I(z', t_{\text{ret}})}{r} \hat{z}$$

Complexify:

$$\vec{A}(\vec{r}, t) = \frac{\mu_0}{4\pi} \int_{-\lambda/4}^{\lambda/4} dz' I(z') \frac{e^{i\omega(t - \frac{r}{c})}}{r} \hat{z}$$

$$\Rightarrow \vec{A}(\vec{r}) = \frac{\mu_0}{4\pi} \int_{-\lambda/4}^{\lambda/4} I(z') \frac{e^{ikr}}{r} \hat{z}$$

In radiation zone:  $r \approx r - z' \cos \theta$

$$\vec{A}_{\text{rad}}(\vec{r}) = \frac{\mu_0}{4\pi} \int_{-\lambda/4}^{\lambda/4} I(z') \frac{e^{ikr \cos \theta}}{r} e^{ikr} \hat{z}$$

Thus:  $\vec{E}_{\text{rad}} = i\omega \vec{A}_{\perp}^{\text{rad}}$ ,  $\hat{z}_{\perp} = \sin\theta \hat{r}$

$$= \frac{\mu_0}{4\pi} i\omega \int_{-\lambda/4}^{\lambda/4} I(z') e^{-ikz'\cos\theta} \frac{e^{ikr}}{r} \sin\theta \hat{r}$$

$$\Rightarrow \boxed{\vec{E}_{\text{rad}} = \frac{1}{4\pi\epsilon_0} \frac{i k \sin\theta}{c} e^{ikr} \frac{1}{r} \int_{-\lambda/4}^{\lambda/4} I(z') e^{-ikz'\cos\theta} dz'}$$

Having used  $\mu_0\epsilon_0 = \frac{1}{c^2}$   $k = \frac{\omega}{c}$

(c) We must do the integral

$$d = \int_{-\lambda/4}^{\lambda/4} I(z') e^{-ikz'\cos\theta} dz' = \int_{-\lambda/4}^{\lambda/4} I_0 \cos\left(\frac{2\pi z'}{\lambda}\right) e^{-ikz'\cos\theta} dz'$$

$$\downarrow$$

$$\frac{e^{ikz'} - e^{-ikz'}}{2}$$

After a little algebra

$$d = \frac{2}{k} \frac{\cos\left(\frac{\pi}{2} \cos\theta\right)}{\sin^2\theta}$$

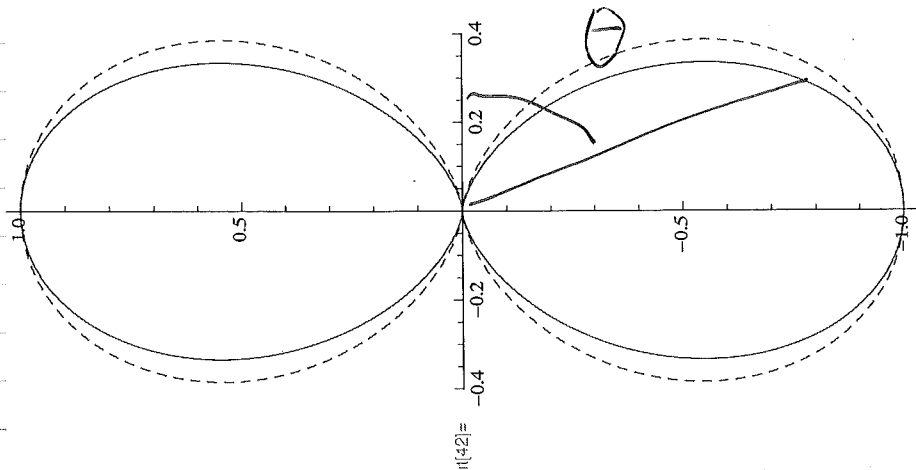
$$\Rightarrow \boxed{\vec{E}_{\text{rad}} = \frac{1}{4\pi\epsilon_0} \frac{2i I_0}{c} \frac{e^{ikr}}{r} \frac{\cos\left(\frac{\pi}{2} \cos\theta\right)}{\sin^2\theta}}$$

This is equivalent to the form given in the assignment with  $l = \frac{\lambda}{2}$

(d) The time-averaged Power squared / solid angle

$$\frac{dP_{\text{rad}}}{d\Omega} \propto \left| \vec{E}_{\text{rad}} \right|^2$$

$$\propto \frac{\cos^2\left(\frac{\pi}{2} \cos\theta\right)}{\sin^2\theta}$$



Above is a polar plot made with Mathematica. The dashed line is the standard dipole pattern  $\sin^2\theta$ . We see that the half-wave antenna concentrates the power into a ~~the~~ higher beam @  $\theta = 90^\circ$ .

(e) We now want to integrate the radiated power over all solid angles to get the total

$$P_{\text{rad}} = \int d\Omega \frac{dP_{\text{rad}}}{d\Omega} = \int 2\pi d(\cos\theta) \frac{dP_{\text{rad}}(\cos\theta)}{d\Omega}$$

$$= 2\pi \int_{-1}^1 d\mu \frac{dP_{\text{rad}}(\mu)}{d\Omega} \quad \text{where } \mu = \cos\theta$$

$$\frac{dP_{\text{rad}}}{d\Omega} = \langle \vec{S} \rangle \cdot \hat{r} r^2 = \frac{c\epsilon_0}{2} |\vec{E}_{\text{rad}}|^2 r^2$$

$$= \frac{I_0^2}{8\pi^2 \epsilon_0} \frac{\cos^2\left(\frac{\pi}{2} \cos\theta\right)}{\sin^2\theta}$$

$$\Rightarrow P_{\text{rad}} = \int \frac{dP_{\text{rad}}}{d\Omega} = \frac{1}{8\pi} Z_0 I_0^2 \int_{-1}^1 2\pi d\mu \frac{\cos^2\left(\frac{\pi}{2} \mu\right)}{1-\mu^2}$$

$$= \frac{1}{4\pi} Z_0 I_0^2 \int_{-1}^1 d\mu \frac{\cos^2\left(\frac{\pi}{2} \mu\right)}{1-\mu^2} = \frac{1}{2} I R_{\text{rad}}^2$$

Numerically  $\approx 1.22$

$$\Rightarrow R_{\text{rad}} = \frac{0.61}{\pi} Z_0 \approx 73 \text{ ohms}$$

To impedance match, this must be equal to the impedance of the coax.

In comparison, in part (a) an ideal dipole antenna has a radiation resistance

$$R_{\text{rad}}^{(\text{dipole})} = \frac{2}{3} \pi Z_0 \left(\frac{d}{\lambda}\right)^2 = \frac{\pi}{6} Z_0 \left(\text{for } d = \frac{\lambda}{2}\right)$$

$$\Rightarrow 197 \text{ ohms} \quad (\text{more than a factor of 2 different})$$